

# Advanced Control of a Quadrotor using Eigenaxis Rotation

Reorientation using LQ-optimal controller with quaternion  
feedback

Jaromir Dvorak    Marcelo de Lellis    Zdenek Hurak

Faculty of Electrical Engineering  
Czech Technical University in Prague  
Prague, Czech Republic

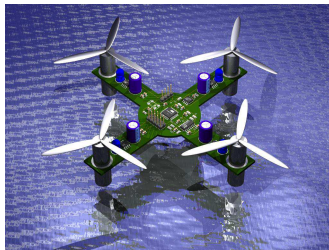
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# Outline

- 1 Problem formulation and practical motivation
- 2 Dynamic model

# Practical motivation

UAVs being developed at our institute

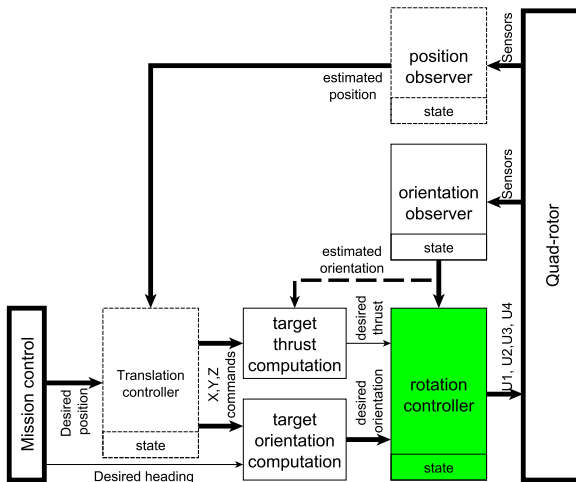


Aiming to synthesize universal rotation (re-orientation) controller

- analytically using the precise dynamic model
- with no path constraints or singularities
- energy efficient with fast settling times
- able to handle large maneuvers

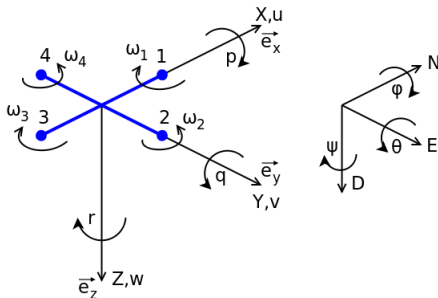
# Controller structure

Prioritary dynamics of the rotation over the translation subsystem



# Problem definition

Reorientation of a rigid body

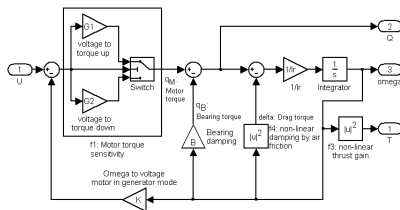


Euler angles vs. Quaternions

# Rotation subsystem dynamic model

3DOF rotation subsystem: 9<sup>th</sup> order non-linear system

- Actuator model



- Rigid body model: acting moments
  - Difference axial thrust and Drag torque from rotors
  - Gyroscopic moments from rotors
  - Aerodynamic resistance moments

# Eigenaxis of rotation

Principal axis of re-orientation maneuver from state  $\mathbf{Q}$  to  $\hat{\mathbf{Q}}$

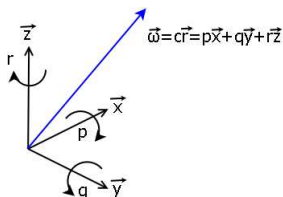
$$\mathbf{E} = \ln(\mathbf{Q}^{-1} \otimes \hat{\mathbf{Q}}) \quad (1)$$

- $\mathbf{E}$  is the eigenaxis vector in body-frame  $\mathbf{Q}$ .
- To reach the target orientation  $\hat{\mathbf{Q}}$ ,  $\mathbf{E}$  must converge to zero.

Quaternion representation offers simple eigenaxis extraction.

# Eigenaxis reorientation

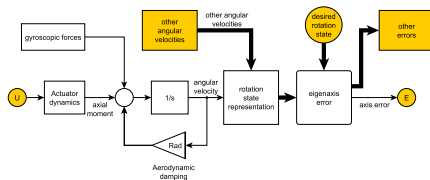
Body rotates around an axis  $\mathbf{r}$  iff its angular velocity  $\omega = c\mathbf{r}$ .



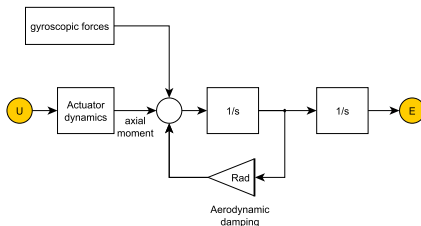
Substituting  $\mathbf{r} = \mathbf{E}$ , reorientation in time  $T$  is performed when  $\omega = c\mathbf{E}$ , where  $\int_0^T c(t)dt = |\mathbf{E}|$ . Such reorientation is called *Eigenaxis reorientation*.

# Dynamic model with eigenaxis output

The non-linear single-axis decoupled dynamics is shown below



Assuming  $\omega = cE$  during  $T$ , model can be linearized



# Problem transformation

$9^{th}$  order non-linear system decoupled to three  $3^{rd}$  order linear systems

- Constraint: equal step responses - continuous recalculation
- LQ-optimal controller with Gyroscopic compensator

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Advantages over the controller using Euler angles decoupling

- no operating points, no singularities
- velocity corresponds to derivative of eigenaxis error
- spherical interpolation, shortest path

# Visualisation

(Loading Circle-m-increase3.mp4)

# Controller synthesis

Task: drive the output  $E$  of the system

$$\begin{aligned}\ddot{E}_x &= \frac{l_a}{I_x} (T_4 - T_2) + \frac{r_{adp}}{I_x} \dot{E}_x + \frac{g_x(\omega, \mathbf{\Omega})}{I_x} \\ \ddot{E}_y &= \frac{l_a}{I_y} (T_1 - T_3) + \frac{r_{adq}}{I_y} \dot{E}_y + \frac{g_y(\omega, \mathbf{\Omega})}{I_y} \\ \ddot{E}_z &= \frac{1}{I_z} (q_1 - q_2 + q_3 - q_4) + \frac{r_{adr}}{I_z} \dot{E}_z\end{aligned}\quad (2)$$

where actuator outputs  $T_j$  and  $q_j$

$$\begin{aligned}l_r \dot{\omega}_j &= k (\Delta u_j) - (m + l) \Delta \omega_j \\ q_j &= l_r \dot{\omega}_j + m \Delta \omega_j + f_4(\omega_0) \\ T_j &= n \Delta \omega_j + f_3(\omega_0)\end{aligned}\quad (3)$$

to zero using the four inputs  $\Delta u_j$ .

# LQ-optimal controller

Model for each axis in matrix form

$$\dot{\mathbf{x}}_j = \begin{bmatrix} -\frac{l+m}{l_r} & 0 & 0 & 0 \\ \frac{nl_a}{l_j} & \frac{r_{adj}}{l_j} & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \mathbf{x}_j + \begin{bmatrix} \frac{k}{l_r} \\ 0 \\ 0 \\ 0 \end{bmatrix} (\tilde{u}_j) \quad (4)$$

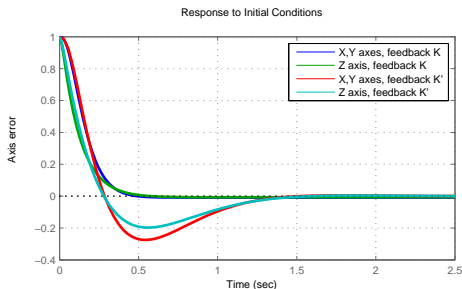
$$E_j = [ 0 \ 0 \ 1 \ 0 ] \mathbf{x}_j$$

- input is the voltage difference for each axis
- additional integrator for asymptotic tracking
- LQ-optimal controller

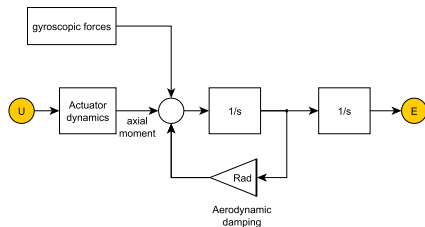
# LQ-optimal controller

State variables for feedback

- $\mathbf{E}$  - computed from  $\mathbf{E} = \ln(\mathbf{Q}^{-1} \otimes \hat{\mathbf{Q}})$
- $\int \mathbf{E}$  - computed using discrete integration
- $\Omega$  - measured by gyro
- $\Delta\omega$  - measured by the BLDC drive



# Gyroscopic compensation



Gyroscopic forces can be computed from  $\Omega$ ,  $\omega$

- Feedforward PD control

# Overall control law

Output mapping

$$\begin{bmatrix} u_1 \\ u_2 \\ u_3 \\ u_4 \end{bmatrix} = \begin{bmatrix} 1 & 0 & -1/2 & 1/4 \\ 1 & 1/2 & 0 & -1/4 \\ 1 & 0 & 1/2 & 1/4 \\ 1 & -1/2 & 0 & -1/4 \end{bmatrix} \begin{bmatrix} \tau \tilde{u} \\ L \tilde{u}_x + G \tilde{u}_x \\ L \tilde{u}_y + G \tilde{u}_y \\ L \tilde{u}_z \end{bmatrix} \quad (5)$$

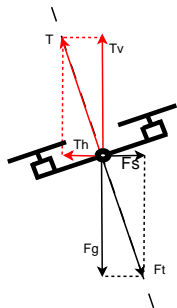
Possible extensions

- overshoot compensation
- inverse non-linearities for the actuator

## Interface to precedent layer

Compute target orientation  $\hat{\mathbf{Q}}$  from

- desired NED acceleration  $\hat{\mathbf{a}}_n$
- desired yaw  $\hat{\psi}$



$$\alpha = \arccos \frac{\hat{\mathbf{a}}_n}{|\hat{\mathbf{a}}_n|} \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} \quad (6)$$

$$\mathbf{Q}_t = \begin{bmatrix} \cos(\frac{\alpha}{2}) \\ \sin(\frac{\alpha}{2}) \frac{\hat{\mathbf{a}}_n}{|\hat{\mathbf{a}}_n|} \times \begin{bmatrix} 0 & 0 & 1 \end{bmatrix}^T \end{bmatrix} \quad (7)$$

$$\mathbf{Q}_y = \begin{bmatrix} \cos(\frac{\hat{\psi}}{2}) & 0 & 0 & \sin(\frac{\hat{\psi}}{2}) \end{bmatrix}^T \quad (8)$$

$$\hat{\mathbf{Q}} = \mathbf{Q}_y \otimes \mathbf{Q}_t \quad (9)$$

# Conclusions, discussions, future

- Simple (intuitive) solutions
- Formalized using two concepts: reset observer and modified Smith compensator
- Bias of the rate sensor leads to deterioration but not instability (integration on finite intervals). Bias estimation techniques needed